

Quantization improves stabilization of dynamical systems with delayed feedback

Gabor Stepan,¹ John G. Milton,² and Tamas Insperger³

¹*Department of Applied Mechanics, Budapest University of Technology and Economics, 1111 Budapest, Hungary*

²*W. M. Keck Science Center, The Claremont Colleges, Claremont, CA 91711, USA*

³*Department of Applied Mechanics, Budapest University of Technology and Economics and MTA-BME Lendület Human Balancing Research Group, 1111 Budapest, Hungary*

(Dated: 17 October 2017)

We show that an unstable scalar dynamical system with time-delayed feedback can be stabilized by quantizing the feedback. The discrete time model corresponds to a previously unrecognized case of the microchaotic map in which the fixed point is both locally and globally repelling. In the continuous-time model, stabilization by quantization is possible when the fixed point in the absence of feedback is an unstable node, and in the presence of feedback it is an unstable focus (spiral). The results are illustrated with numerical simulation of the unstable Hayes equation. The solutions of the quantized Hayes equation take the form of oscillations in which the amplitude is a function of the size of the quantization step. If the quantization step is sufficiently small, the amplitude of the oscillations can be small enough to practically approximate the dynamics around a stable fixed point.

Digital feedback controllers are having an increasing impact on human activities. Examples range from the control of single-atom trajectories¹ to the development of brain-device interfaces², from the treatment of human diseases using closed-loop drug delivery systems³, to the design of driverless automobiles⁴. An important component of these controllers is an analog-to-digital (A/D) conversion by which a continuous (or analog) signal is converted into a series of numbers proportional to the signal. This A/D conversion is essential to enable the digital microprocessors to track the controlled variable: the reverse D/A conversion makes it possible for the microprocessor to affect control. The effect of A/D and D/A conversion is to introduce a quantization into both the time domain (“sampling”) and the controlling force (“round off”). Although it is well known that these quantizations can introduce spurious oscillations^{5,6} and small amplitude micro-chaotic fluctuations into the control dynamics^{7,8}, less is known about the possible benefits of feedback quantization for control. Here we analyze a generic scalar model for feedback control and show that a “coarse-grained” signal quantization can contribute to the stabilization of unstable dynamical systems in the presence of feedback delays.

I. INTRODUCTION

The advantages of digital feedback control over continuous, or analog, control are well documented^{9,10}. Digital controllers are cheaper, easier to configure, more adaptable and less prone to the effects of environmental fluctuations. However, the dynamics which arise in the setting of digital control can sometimes be counter-intuitive. Examples include the appearance of an oscillation whose frequency is within a range far smaller than the sampling frequency and the appearance of low amplitude stochastic-like fluctuations in the desired motion, referred to as *microchaos*^{11,12}. Even more paradoxical is the prediction that transiently stable solutions can occur for parameter ranges that for analog feedback would be unstable^{13,14}. These quantization effects may not be unique to man-made devices. The conversion of sensory inputs into a train of discrete action potentials underlies sensory-motor encoding in the nervous system. Quantization of voluntary movements are manifested in the visually guided movements in infants¹⁵ and in patients with brain injury¹⁶. Moreover, sensory dead zones, namely a range of inputs which give the same output, arise both in neural control¹⁷ and in A/D conversion^{18,19} and result in the appearance of limit cycle oscillations.

This communication focuses on the observation that certain unstable time-delayed dynamical systems can be practically stabilized by round off. Our discussion is organized as follows. In Section II, we review the effects of feedback digitization on the dynamics of time delayed feedback control from the perspective of a discrete time map referred to as the *microchaos map*. Our focus is on the stable dynamics which arise when the fixed-point is locally unstable (repelling). Section III demonstrates

that the stable dynamics that arise when this fixed point is globally attracting include limit cycle oscillations and microchaos. Surprisingly, stable and metastable oscillatory dynamics can also arise if this fixed point in the presence of continuous feedback is, in addition, globally repelling. Section IV shows an application of these observations for stabilizing the Hayes equation⁷ for delayed feedback control using quantization. Since the amplitude of the fluctuations is proportional to the size of the quantization step, it becomes possible to replace the dynamics around an unstable fixed point with stable, low amplitude fluctuations. This result may be sufficient for many practical applications.

II. DIGITAL FEEDBACK

Consider the dynamics of an over-damped unstable system described by

$$\dot{x}(t) = qx(t) - f(x(t)), \quad (1)$$

where x is a state variable, $q > 0$ is a constant and f describes the feedback. There are two effects of a digital implementation of (1). First, the fact that f is determined from the sampled values of x introduces a temporal piecewise smooth dynamics^{7, 8} governed by

$$\dot{x}(t) = qx(t) - f(x(t_j)), \quad t \in [t_j, t_{j+1}), \quad (2)$$

where $t_j = j\Delta t$, $j = 0, 1, 2, \dots$ are the sampling instants and Δt is the constant sampling time. Temporal sampling introduces a delay in the system, hence Eq. (2) can also be written as

$$\dot{x}(t) = qx(t) - f(x(t - \rho(t))), \quad (3)$$

where

$$\rho(t) = t - \Delta t \text{Int} \left(\frac{t}{\Delta t} \right) \quad (4)$$

is a time-periodic delay, and the function $\text{Int}()$ rounds towards zero. The average delay, $\bar{\rho}$, is

$$\bar{\rho} = \frac{1}{\Delta t} \int_0^{\Delta t} \rho(t) dt = \frac{\Delta t}{2}. \quad (5)$$

Second, the quantization round off of f means that the feedback forces are computed using integer multiples of the quantization step, h , and hence

$$f(x(t_j)) = ph \text{Int} \left(\frac{x(t_j)}{h} \right), \quad (6)$$

where we have assumed linear state feedback with control gain p . In making the above approximations, we assume a zero-order hold, namely the force is kept constant over the interval Δt . A consequence of this assumption is that despite the periodically varying delay, the delayed feedback becomes a piecewise constant function.

III. DISCRETE TIME: MICROCHAOTIC MAP

Equation (3) with (4) and (6) leads to the governing equation

$$\dot{x}(t) = qx(t) - ph \text{Int} \left(\frac{1}{h} x \left(t - \Delta t \text{Int} \left(\frac{1}{\Delta t} t \right) \right) \right) \quad (7)$$

in continuous-time representation. The internal Int function refers to discretization in time, the external Int function represents quantization in space. Discretization in time is a linear effect that increases the dimension of the state space due to the inherent delay; quantization in space makes the problem strongly nonlinear.

Equation (2) with (6) gives an equivalent semi-discrete form

$$\dot{x}(t) = qx(t) - ph \text{Int} \left(\frac{x(t_j)}{h} \right), \quad t \in [t_j, t_{j+1}). \quad (8)$$

This equation can be solved over the interval $[t_j, t_{j+1})$ to give the discrete governing equation in the form of the *microchaos map*⁷

$$x(t_{j+1}) = ax(t_j) - bh \text{Int} \left(\frac{x(t_j)}{h} \right), \quad (9)$$

where

$$a = \exp(q\Delta t) > 1, \quad b = \frac{p}{q}(1 - \exp(q\Delta t)). \quad (10)$$

The dynamics of (9) arise from the interplay between behaviors close to the trivial fixed point $x \equiv 0$ and those global to it. The local dynamics, when $|x| < h$, is governed by

$$x(t_{j+1}) = ax(t_j). \quad (11)$$

Since $a > 1$, the trivial fixed point is always locally ‘‘cyclic fold’’ unstable. The phrase ‘‘cyclic fold’’ means that the changes in x occur monotonically. When $|x| \gg h$, then $bh \text{Int} \left(\frac{x(t_j)}{h} \right) \approx bx(t_j)$ and hence the global dynamics are governed by

$$x(t_{j+1}) = (a - b)x(t_j). \quad (12)$$

With respect to the global dynamics, four different cases (labeled respectively, A, B, C and D) can be distinguished depending on the relation between a and b . Local and global behavior of the different cases are illustrated in Figure 1. Different regions in the parameter plane (a, b) are shown in Figure 2.

In what follows, we show that three out of the four cases, namely, Cases B, C and D, present microchaos. The three conditions for chaos we use here are⁷: (1) sensitive dependence on the initial conditions; (2) existence of closed invariant attractive sets; and (3) topological transitivity (mixing). First, the sensitive dependence on the initial conditions follows directly from the positiveness of the Lyapunov exponent $\Lambda = \ln(a) > 0$ for

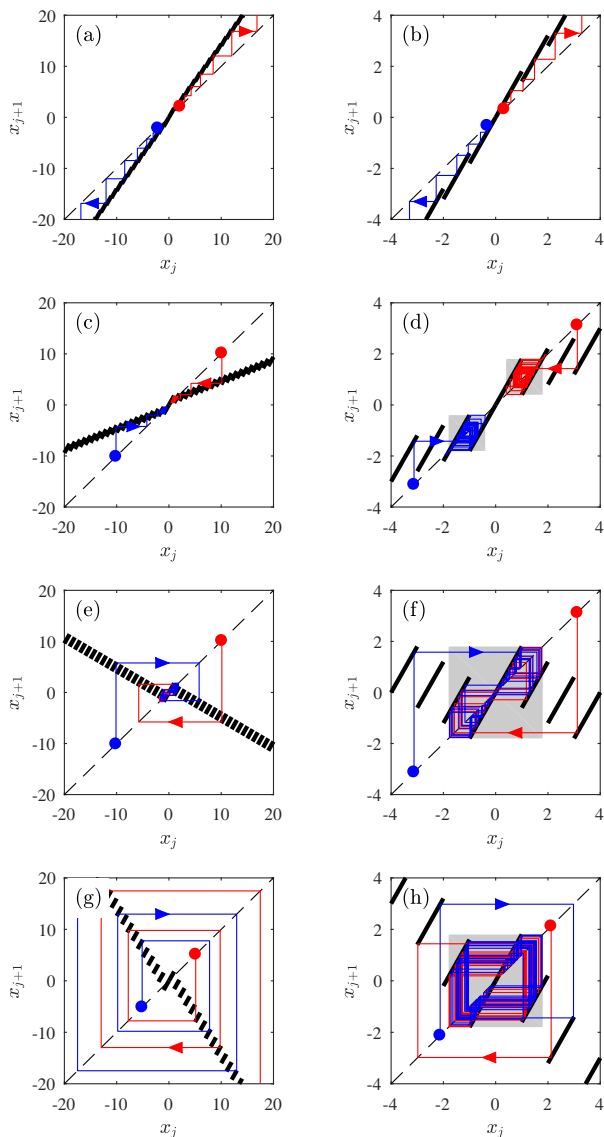


FIG. 1. Cobweb diagrams for (9) with $h = 1$ when $a = 1.8$ and (a,b) $b = 0.4$, (c,d) $b = 1.4$, (e,f) $b = 2.4$, (g,h) $b = 3.4$. The red and blue lines shows different solutions associated with two different choices of the initial conditions denoted by red and blue dots, respectively. The thick black lines represent the right-hand side of (9). Left panels represent global dynamics, while right panels illustrate the dynamics closer to the trivial fixed point.

all of the cases. Second, closed invariant attractive sets, which confine chaotic motions, are obtained. Third, the mixing property is not proven here, but we conjecture that existing methods based on the construction of an appropriate symbolic dynamics[?] can be extended to establish topological transitivity of these maps.

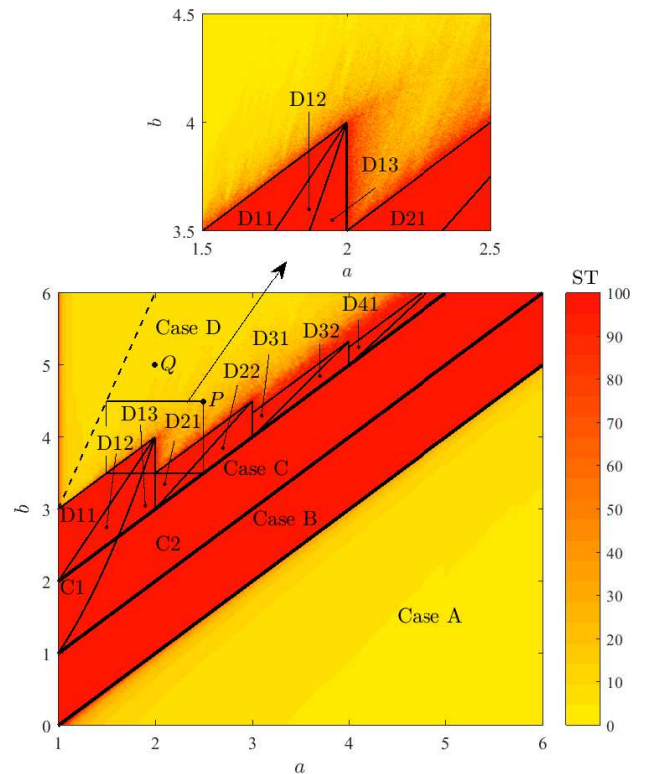


FIG. 2. Steady state behavior of (9). Yellow-red color code indicates the expected survival time (ST) determined via a series of numerical simulations. Parameter regions for Cases A, B, C and D are separated by thick lines, while individual sub regions associated with different size of domains of attractions are indicated by thin lines. The dashed line is $b = 3a$ (see text for discussion).

A. Case A: $a - b > 1$

The trivial fixed point is globally cyclic fold repelling (see Figure 1a). Since this case admits no sustained solution we do not consider it further.

B. Case B: $0 < a - b < 1$

This case has been analyzed previously by Haller and Stepan[?]. The neighborhood of the trivial fixed point is globally cyclic fold attracting (Figure 1c). Note that for sufficiently large x_j (“large scale”) the values of the iterates decrease (Figure 1c). However since the fixed point $x \equiv 0$ is unstable, the dynamics increase for small x_j (“small scale”) (Figure 1d). For certain parameter combinations it can be proved that the dynamics are microchaotic[?]. However, the numerical results in Figure 2 suggest that microchaotic solutions exist at each point of the parameter region $0 < a - b < 1$. The closed invariant attractive set confining microchaotic solutions

is

$$\mathcal{A}_{B+} = [n(a-b)h, (m(a-b)+a)h], \quad (13)$$

for positive initial conditions[?] and

$$\mathcal{A}_{B-} = [-(m(a-b)+a)h, -n(a-b)h], \quad (14)$$

for negative initial conditions, where

$$n = \text{floor} \left(\frac{b}{(b+1-a)a} \right), \quad m = \text{ceil} \left(\frac{(a-1)a-b}{(b+1-a)a} \right).$$

Note that the size of the invariant attractive set is scaled to the quantization step h . That is, for small quantization step, the amplitude of the chaotic motions is small. The fact that the chaotic dynamics are confined to a small region near the trivial fixed point justifies the term microchaos. The domain of attraction of \mathcal{A}_{B+} is $\mathcal{D}_{B+} = (0, \infty)$, while the domain of attraction of \mathcal{A}_{B-} is $\mathcal{D}_{B-} = (-\infty, 0)$.

C. Case C: $-1 < a - b < 0$

The neighborhood of the trivial fixed point is globally period doubling attracting (Figure 1e). The phrase ‘‘period doubling’’ means that the changes in x occur with alternating sign. Although this particular case has not been analyzed previously, numerical simulations indicate that these solutions are microchaotic. The closed invariant attractive set for Case C is

$$\mathcal{A}_{C1} = [-ah, ah] \setminus [(a^2 - b)h, (b - a^2)h], \quad (15)$$

when $b > a^2$ (see region C1 in Figure 2) and

$$\mathcal{A}_{C2} = [-ah, ah], \quad (16)$$

when $b \leq a^2$ (see region C2 in Figure 2). For details on the calculation of \mathcal{A}_{C1} and \mathcal{A}_{C2} , see Appendix A. Similarly to Case B, the size of the invariant attractive set has been scaled to the quantization step h . The domain of attraction of \mathcal{A}_{C1} and \mathcal{A}_{C2} is $\mathcal{D}_C = (-\infty, \infty)$.

D. Case D: $a - b < -1$

The neighborhood of the trivial fixed point is globally period doubling repelling (Figure 1g). This case has not been studied previously. Despite the fact that the fixed point is both locally and globally repelling, it is possible that a stable microchaotic solution exists (Figure 1h). Numerical simulations show that the regions D11, D12, D13, D21, D22 etc. in Figure 2 are associated with permanent chaos. The boundaries of these parameter domains and the corresponding closed invariant attractive

sets can be given as

$$\begin{aligned} \text{D11: } & 1 < a < 2 \text{ and } 2a < b < a + 2, \\ & \mathcal{A}_{\text{D11}} = [(a-b)h, (b-a)h] \\ & \quad \setminus [(a(b-a)-b)h, (b-a(b-a))h], \end{aligned} \quad (17)$$

$$\begin{aligned} \text{D12: } & 1 < a < 2 \text{ and } a^2 < b < 2a \text{ and } a + 1 < b, \\ & \mathcal{A}_{\text{D12}} = [-ah, ah] \setminus [(a^2 - b)h, (b - a^2)h], \end{aligned} \quad (18)$$

$$\begin{aligned} \text{D13: } & 1 < a < 2 \text{ and } a + 1 < b < a^2, \\ & \mathcal{A}_{\text{D13}} = [-ah, ah], \end{aligned} \quad (19)$$

$$\begin{aligned} \text{D21: } & 2 < a < 3 \text{ and } \frac{3}{2}a < b < a + \frac{3}{2}, \\ & \mathcal{A}_{\text{D21}} = [2(a-b)h, 2(b-a)h], \end{aligned} \quad (20)$$

$$\begin{aligned} \text{D22: } & 2 < a < 3 \text{ and } a + 1 < b < \frac{3}{2}a, \\ & \mathcal{A}_{\text{D22}} = [-ah, ah], \end{aligned} \quad (21)$$

$$\begin{aligned} \text{D31: } & 3 < a < 4 \text{ and } \frac{4}{3}a < b < a + \frac{4}{3}, \\ & \mathcal{A}_{\text{D31}} = [3(a-b)h, 3(b-a)h], \end{aligned} \quad (22)$$

$$\begin{aligned} \text{D32: } & 3 < a < 4 \text{ and } a + 1 < b < \frac{4}{3}a, \\ & \mathcal{A}_{\text{D32}} = [-ah, ah], \end{aligned} \quad (23)$$

...

(see Appendix A for details). Similar to Cases B and C, the size of the invariant attractive set is scaled to the quantization step h . As opposed to Cases B and C, the domain of attraction of the above attractive sets is not infinite, since the global dynamics (for large $|x|$) in Case D is repelling. This means that for small quantization step, the solution may easily escape from the invariant attractive set due to large enough disturbances or do not even get inside this set if the initial condition is large enough (see the solutions in Figure 1c,e,g for Cases B, C and D).

E. Transient microchaos

In contrast to Cases B and C, transient microchaotic dynamics can also arise in Case D. The term ‘‘transient microchaos’’ refers to metastable solutions that transiently survive close to the trivial equilibrium before diverging towards infinity[?]. These transient solutions exist just above the regions D11, D21, D31, etc. and below the line $b = 3a$ indicated by dashed line in Figure 2. Transient microchaotic dynamics can be characterized by the first passage or survival time (ST), namely, the time at which $|x|$ first exceeds a certain value, x_{lim} . An added complexity is the fact that the ST is sensitive to the choice of initial condition (see Figure 3). In general, the expected value of the ST cannot be determined analytically except for special cases (see Appendix B for the determination of the ST for point Q in Figure 2).

In view of the above considerations, the ST in Figure 2 were determined numerically as follows. Time history for a series of pairs (a, b) was determined numerically for 20 different initial conditions distributed uniformly in the interval $[-h, h]$ with $h = 1$ for $t_{\text{max}} = 100$ iteration steps

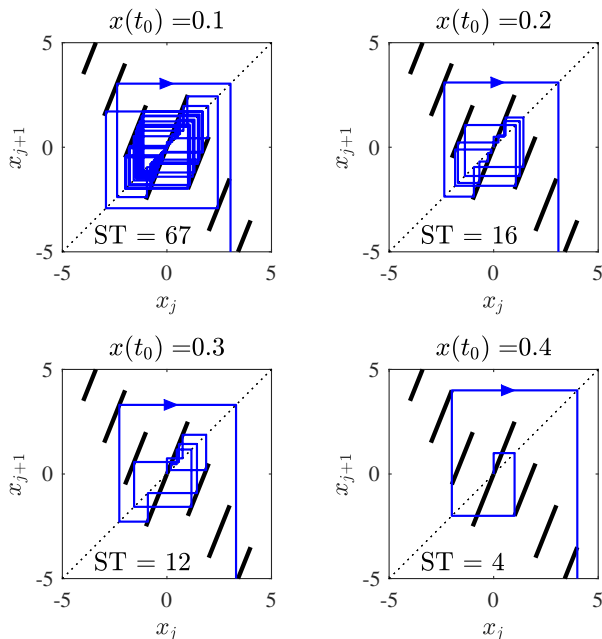


FIG. 3. Cobweb diagrams for Case D with $h = 1$ associated with point P ($a = 2.5$, $b = 4.5$) in Figure 2 with different initial conditions $x(t_0)$.

and the time instant where $|x|$ first exceeded $x_{\text{lim}} = 100$ was recorded and averaged. The color code indicates the averaged ST: yellow indicates $ST = 0$ and red indicates $ST = t_{\text{max}}$.

It is observed that the ST in the region associated with Case A is close to zero as expected, since both the open-loop and the closed loop system is unstable. The regions of Case B and Case C are both associated with red color. The ST in these regions is $t_{\text{max}} = 100$, which reflects the existence of long-lived bounded motions, namely, a permanent microchaos as shown in Figures 1d and 1f, respectively. For Case D there are choices of (a, b) which are associated with long-lived bounded microchaos (red) as well as parameter choices with transient solutions (red-yellow scale).

IV. QUANTIZATION: HAYES EQUATION

We anticipate that the behaviors exhibited by Case D will have their counterpart in continuous time-delayed feedback control systems. Of particular interest is the possibility of stabilization through quantization of the feedback. To explore this possibility we consider the Hayes equation[?]

$$\dot{x}(t) = qx(t) - px(t - T), \quad (24)$$

where $q > 0$ is the system parameter, p is the control gain and T is the feedback delay. By rescaling time as $t := qt$ and dropping the tilde immediately we get

$$\dot{x}(t) = x(t) - Cx(t - \tau). \quad (25)$$

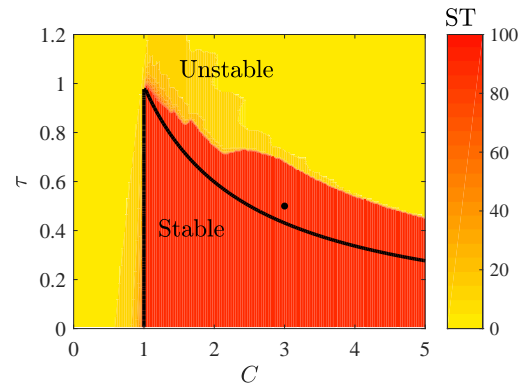


FIG. 4. Regions of steady state behaviors of equations (25) and (27) when $h = 1$. Solid lines indicate the stability boundaries for the Hayes equation (25). Yellow-red color code indicates the expected survival time (ST) for the quantized Hayes equation (25). The \bullet indicates the point $C = 3$, $\tau = 0.5$ discussed in Figure 5.

where $\tau = T/q$ is the scaled feedback delay and $C = p/q > 0$ is the scaled feedback gain. Biological applications of the Hayes equation arise in the description of the control of blood cell dynamics[?] and the pupil light reflex[?]. When $C = 0$, the fixed point $x \equiv 0$ is an unstable node. When τ and C are sufficiently large stability is lost and we have an unstable spiral point. The stability boundaries of (25) in the plane (C, τ) are given by the line $C = 1$ and the parametric curve

$$C = \sqrt{\omega^2 + 1}, \quad \tau = \frac{1}{\omega} \text{atan}(\omega) \quad (26)$$

with $\omega \in [0, \infty)$. Quantization of (25) yields

$$\dot{x}(t) = x(t) - Ch \text{Int} \left(\frac{x(t - \tau)}{h} \right), \quad (27)$$

where h is the quantization step. When $|x| \gg h$, then $Ch \text{Int} \left(\frac{x(t - \tau)}{h} \right) \approx Cx(t - \tau)$ and hence the global dynamics are still governed by (25). Alternatively, one can write (27) in the form

$$\dot{x}(t) = \begin{cases} \dots \\ x(t) - 2Ch & \text{if } 2h \leq x(t - \tau) < 3h, \\ x(t) - Ch & \text{if } h \leq x(t - \tau) < 2h, \\ x(t) & \text{if } -h < x(t - \tau) < h, \\ x(t) + Ch & \text{if } -2h < x(t - \tau) \leq -h, \\ x(t) + 2Ch & \text{if } -3h < x(t - \tau) \leq -2h, \\ \dots \end{cases}$$

Figure 4 summarizes the behavior of (27) as a function of C and τ when $h = 1$. The semi-discretization numerical method was used to get the time history of the systems for different pairs of (C, τ) over the period $[0, t_{\text{max}}]$ with $t_{\text{max}} = 100$. The initial condition for the simulations was $x(\theta) \equiv x_{\text{IC}}$, $\theta \in [-\tau, 0]$, where x_{IC} is a constant

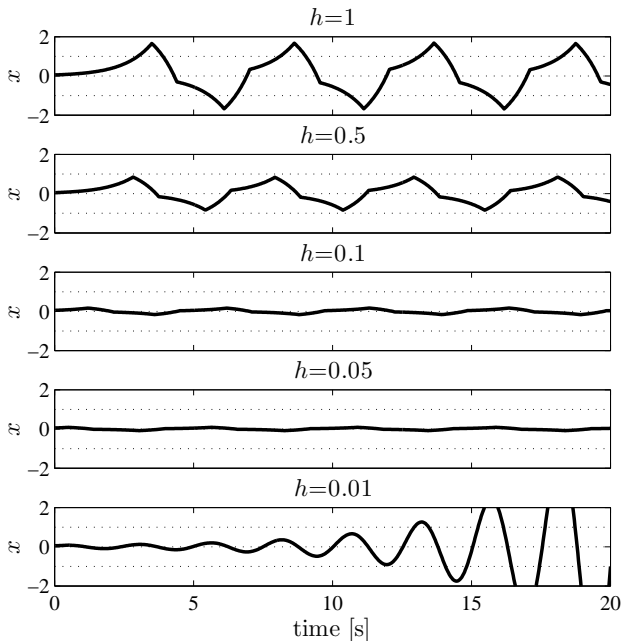


FIG. 5. Effect of quantization step size, h , on the solutions of (27) when $x_{IC} = 0.01$, $C = 3$, $\tau = 0.5$.

chosen such that $x_{IC} \leq h$. Solutions were declared to diverge if $|x(t)| \geq x_{lim} = 100$ was satisfied. Yellow-red color code indicates the ST, i.e., the time when $|x(t)|$ exceeded x_{lim} . It can be seen that bounded motions exist in the region where the Hayes equation (25) is unstable.

Figure 5 shows the effects of changing h on the solutions of (27) when the parameters C, τ are chosen so that the fixed point of (25) is unstable (see \bullet in Figure 4). As can be seen the amplitude of the oscillations can be reduced by decreasing h . However, for cases when $h < x_{IC}$, the solution gets out of the domain of attraction and its amplitude grows exponentially. Thus, there is a trade-off between h and x_{IC} (or between h and the noise in the system). Decreasing h may provide a solution which is a useful approximation for a stabilized fixed point for practical purposes (see the case $h = 0.05$ in Figure 5), however, further decrease of h may destabilize the system.

V. DISCUSSION

Here we have investigated the potential for “coarse grained” quantized feedback to stabilize an unstable feedback control system. We have shown that even when the fixed point is both locally and globally unstable feedback quantization can produce a long-lived bounded solution. This stabilized solution takes the form of an oscillatory fluctuation whose amplitude is proportional to the quantization step h . We did not directly address whether the combination of feedback quantization and time discretization could generate microchaotic so-

lutions as proven for the closely related Eurich-Milton equation[?].

One way that quantization can enter into the dynamics of time delayed feedback control is because of the presence of a sensory dead zones. For example, for stick balancing on the fingertip, there is a sensory dead zone of $1 - 3^\circ$ for the detection of vertical displacement angle in the anterior-posterior direction[?]. Our observations suggests that the presence of this dead zone may not simply be a limitation for control, but could be beneficial.

For the discrete-time system (9), the regions near the trivial fixed point where bounded solutions exist is located at higher gain values when $b > a + 1$, while the system is unstable when $b < a - 1$ (see Figure 2). Similarly, for the quantized Hayes equation (27), bounded solutions exist at higher gain values when C is larger than the stability boundary of the Hayes equation (25), while the system is unstable when $C < 1$ (see Figure 4). This shows that the beneficial effect of quantization can be utilized when the controller is tuned to “overreact” to the changes in the state variable.

Time discretization necessarily introduces a time delay into the dynamical system. This is because the state of the dynamical system at time $t \in [t_j, t_{j+1}]$, $t_j = j\Delta t$ depends on its state at time t_j , where Δt is the discretization step. Practical experience suggests that since the time discretization step can be made very small, it can be ignored; however, this is not always true[?]. A major exception occurs when the feedback itself is time-delayed and quantized[?]. Historically the effect of piecewise constant, time-delayed feedback was extensively studied in early investigations into the dynamics of time-delayed feedback control[?]. Since experimental paradigms could be readily developed, it was possible to directly compare prediction with observation[?]. It is important to note that in these models the feedback switching times could be precisely computed analytically and thus the solutions were obtained by piecing together exponential segments or spiral arcs. The use of numerical algorithms, such as the Euler-discretization, semi-discretization or Runge-Kutta method introduces a low amplitude, microchaotic element to the dynamics[?]. A related situation likely occurs in the use of computer algorithms to simulate the dynamics of integrate-and-fire neurons[?], although in this case a formal demonstration that the dynamics are microchaotic has not yet been made.

Quantizing feedback is expected to be an effective stabilization strategy provided that 1) the fixed point of the uncontrolled dynamical system is exponentially unstable, 2) the fixed point in the presence of continuous delayed feedback is an unstable spiral point, and 3) the noise intensity is not too high. A consequence of time-delayed, quantized feedback is that a both open-loop and closed-loop unstable fixed point is replaced by a stable oscillation. If, in addition, there is time discretization, then the oscillations will be replaced by a microchaotic fluctuation. If the amplitude of the generated oscillation is not

too large then the solution may be acceptable for certain applications. Indeed, the long lived balanced state obtained by experts who balance a stick on their fingertip is a transient microchaotic solution⁷. Another application of these results is that the control gain can be increased by increasing the size of the quantization step, which can be a useful feature for instance in position control in the presence of dry friction.

ACKNOWLEDGEMENT

The research has received funding from the European Research Council under the European Unions Seventh Framework Programme (FP7/2007-2013) / ERC Advanced grant agreement No340889 (SG,IT) and from the William R Kenan, Jr Charitable trust (JM).

Appendix A: Closed invariant sets for Cases B, C and D

The closed invariant attractive sets are the sets of minimum size, where the solutions never escape from. These sets are determined by the parameters a and b and the quantization step h .

The closed invariant attractive sets for Case B, when $0 < a - b < 1$, were given by Haller and Stepan⁷.

For Case C, when $-1 < a - b < 0$, there are two different cases. If $a^2 - b > 0$, then then it is easy to see that the solutions never leaves the set \mathcal{A}_{C2} given in (16), see panel C2 in Figure 6. However, when $a^2 - b < 0$, then the solutions never return to the set $[(a^2 - b)h, (b - a^2)h]$ once they leaved it, hence the invariant attractive set is \mathcal{A}_{C1} given in (15), see panel C1 in Figure 6.

For Case D, when $a - b < -1$, there are several sub-cases. Consider first the case when $1 < a < 2$. If $a > b - a$ and $a^2 - b > 0$, then the invariant set is \mathcal{A}_{D13} given by (19), see panel D13 in Figure 6. If $a > b - a$ but $a^2 - b < 0$, then the solutions never return to the set $[(a^2 - b)h, (b - a^2)h]$ once they leaved it, hence the invariant set is \mathcal{A}_{D12} given by (18). If $a < b - a$, then upper bound of the invariant set becomes $b - a$, the solutions never return to the set $[(a(b - a) - b)h, (b - (a(b - a)))h]$ once they leaved it, hence the invariant set is \mathcal{A}_{D11} given by (17).

Consider now the case when $2 < a < 3$. If $a > 2(b - a)$, then the invariant set is \mathcal{A}_{D22} given by (21), see panel D22 in Figure 6. If $a < 2(b - a)$, then the invariant set is \mathcal{A}_{D21} given by (20), see panel D21 in Figure 6.

The invariant sets for the cases when $a > 3$ can be obtained similarly.

Appendix B: Survival time calculation

The survival time (ST) for (9) with $a = 2$ and $b = 5$ is calculated. This parameter combination corresponds

to point Q in Figure 2. Without loss of generality, the quantization step can be set to $h = 1$.

First, we determine the expected ST for the interval $I_0 = [-2, 2]$, i.e., the average time duration until the solutions initiated from the interval I_0 stay in I_0 . For this calculation, I_0 is divided into subintervals, where the ST is invariant. It is easy to see, for instance, that a solution escapes from the interval I_0 after one step of iteration if $x(t_0) \in I_1$, where

$$I_1 = [-1.5, -1) \cup (1, 1.5]. \quad (B1)$$

The solutions escape after two iterations if $x(t_0) \in I_2$, where

$$I_2 = [-2, -1.75) \cup [-1, -0.5) \cup (0.5, 1] \cup (1.75, 2] \quad (B2)$$

is the pre-image of I_1 . The solution escapes after three iterations if $x(t_0) \in I_3$, where

$$I_3 = [-1.625, -1.5) \cup [-1, -0.875) \cup [-0.375, -0.25) \cup (0.25, -0.375] \cup (0.875, 1] \cup (1.5, 1.625] \quad (B3)$$

is the pre-image of I_2 . By continuing the pre-image mapping, an infinite series of subintervals I_k , $k = 1, 2, \dots$ is obtained, each associated with a ST equal to k (see Figure 7). The length of the subintervals are

$$\ell(I_k) = \frac{2k}{2^k} \quad k = 1, 2, \dots \quad (B4)$$

The length of the union of these subintervals is

$$\sum_{k=1}^{\infty} \ell(I_k) = \sum_{k=1}^{\infty} \frac{2k}{2^k} = \sum_{k=1}^{\infty} k \left(\frac{1}{2}\right)^{k-1}, \quad (B5)$$

which can be calculated as follows. Introduce the function

$$f(q) := \sum_{k=1}^{\infty} q^k = \frac{q}{1-q}, \quad (B6)$$

where $0 < q < 1$. The derivative of f with respect to q is

$$f'(q) = \sum_{k=1}^{\infty} kq^{k-1} = \frac{(1-q) + q}{(1-q)^2} = \frac{1}{(1-q)^2}. \quad (B7)$$

The infinite sum in (B5) can be given by setting $q = 1/2$ in (B7) as

$$\sum_{k=1}^{\infty} \ell(I_k) = f'(1/2) = \frac{1}{(1-1/2)^2} = 4. \quad (B8)$$

Thus, the union of the subintervals I_1, I_2, \dots covers the whole interval I_0 .

The average ST for I_0 can be given as

$$\text{ST}_{I_0} = \frac{1}{\ell(I_0)} \sum_{k=1}^{\infty} k \frac{2k}{2^k} = \frac{1}{\ell(I_0)} \sum_{k=1}^{\infty} k^2 \left(\frac{1}{2}\right)^{k-1}. \quad (B9)$$

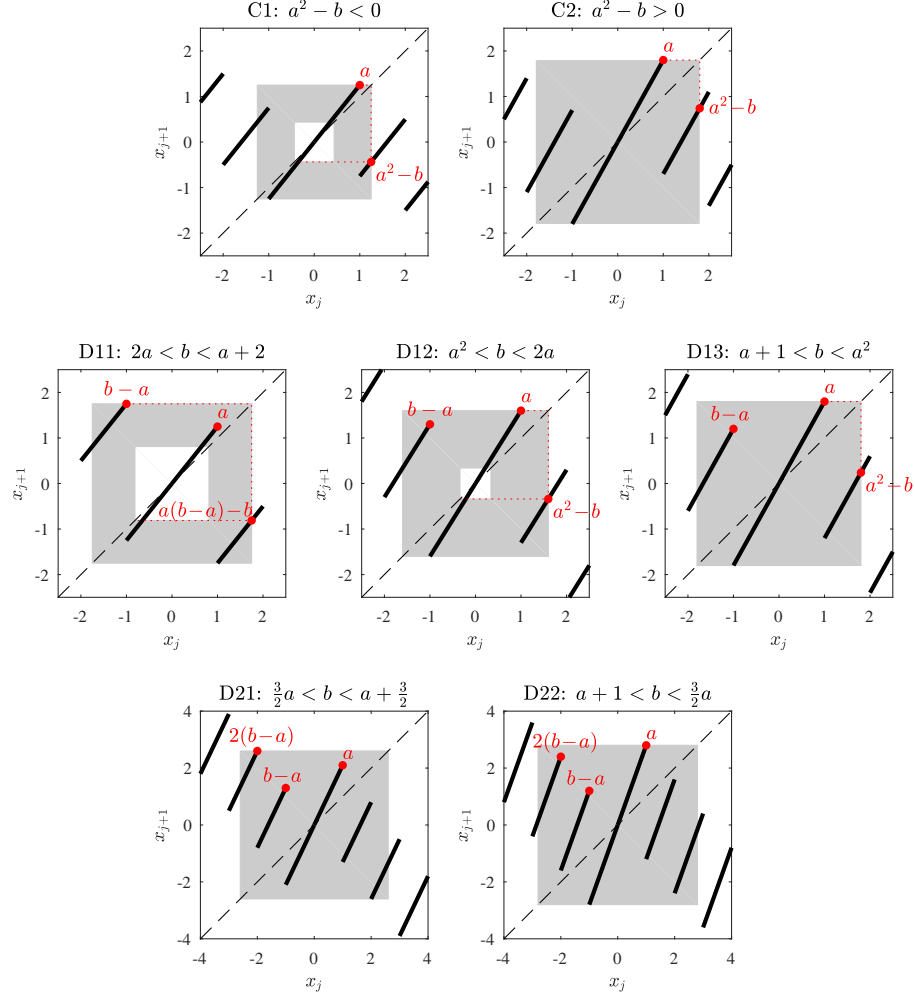


FIG. 6. Illustration of the invariant sets for different cases.

Now introduce the function

$$g(q) := \sum_{k=1}^{\infty} kq^k = q \sum_{k=1}^{\infty} kq^{k-1} = qf'(q) = \frac{q}{(1-q)^2}, \quad (\text{B10})$$

where $0 < q < 1$. The derivative of g with respect to q is

$$g'(q) = \sum_{k=1}^{\infty} k^2 q^{k-1} = \frac{(1-q)^2 + 2q(1-q)}{(1-q)^4} = \frac{1-q^2}{(1-q)^4}. \quad (\text{B11})$$

The infinite sum in (B9) can be given by setting $q = 1/2$ in (B11) as

$$\text{ST}_{I_0} = \frac{1}{\ell(I_0)} g'\left(\frac{1}{2}\right) = \frac{1}{4} \frac{1 - \left(\frac{1}{2}\right)^2}{\left(1 - \frac{1}{2}\right)^4} = \frac{12}{4} = 3. \quad (\text{B12})$$

This is the average ST for the solutions initiated from the interval I_0 . Figure 7 shows a sample solution for a particular initial condition $x(t_0) = 0.00651$ with $\text{ST} = 22$.

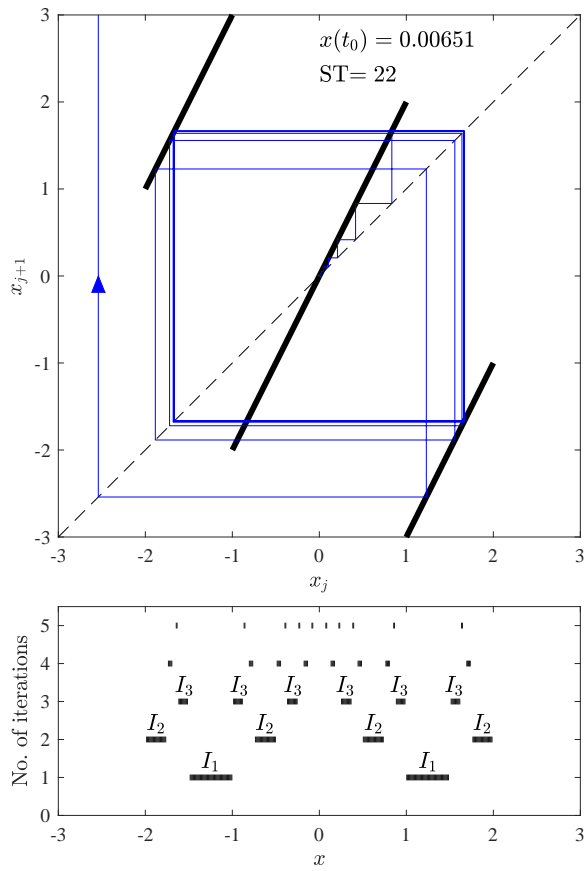


FIG. 7. Cobweb diagram for Case D with $h = 1$ associated with point Q ($a = 2$, $b = 5$) in Figure 2 and the structure of the associated subintervals I_j , $j = 1, 2, \dots$.